# Linear Regulator, Low V<sub>IN</sub>, Low Noise and High PSRR, 200 mA

The NCP110 is a linear regulator capable of supplying 200 mA output current from 1.1 V input voltage. The device provides wide output range from 0.6 V up to 4.0 V, very low noise and high PSRR. Due to low quiescent current the NCP110 is suitable for battery powered devices such as smartphones and tablets. The device is designed to work with a 1 µF input and a 1 µF output ceramic capacitor. It is available in ultra-small 0.35P, 0.64 mm x 0.64 mm Chip Scale Package (CSP) and XDFN4 0.65P, 1 mm x 1 mm.

#### **Features**

- Operating Input Voltage Range: 1.1 V to 5.5 V
- Available in Fixed Voltage Option: 0.6 V to 4.0 V
- ±2% Accuracy Over Load/Temperature
- Ultra Low Quiescent Current Typ. 20 μA
- Standby Current: Typ. 0.1 μA
- Very Low Dropout: 70 mV for 1.05 V @ 100 mA
- High PSRR: Typ. 95 dB at 20 mA, f = 1 kHz
- Ultra Low Noise: 8.8 μV<sub>RMS</sub>
- Stable with a 1 µF Small Case Size Ceramic Capacitors
- Available in -WLCSP4 0.64mm x 0.64mm x 0.33mm Case 567VS -XDFN4 1mm x 1mm x 0.4mm - Case 711AJ
- These Devices are Pb-Free, Halogen Free/BFR Free and are RoHS Compliant

## **Typical Applications**

- Battery-powered Equipment
- Smartphone, Tablets
- Digital Cameras
- Smoke Detectors
- Portable Medical Equipment
- RF, PLL, VCO and Clock Power Supplies
- Battery Powered Wireless IoT Modules

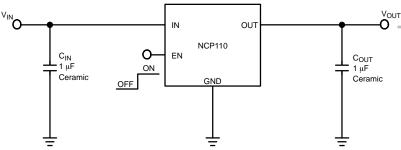


Figure 1. Typical Application Schematics



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## **MARKING DIAGRAMS**



WLCSP4 CASE 567VS



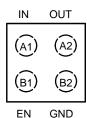


XDFN4 CASE 711AJ

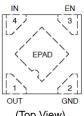


X or XX = Specific Device Code = Date Code

## **PIN CONNECTIONS**



(Top View)



## (Top View)

## **ORDERING INFORMATION**

See detailed ordering, marking and shipping information on page 14 of this data sheet.

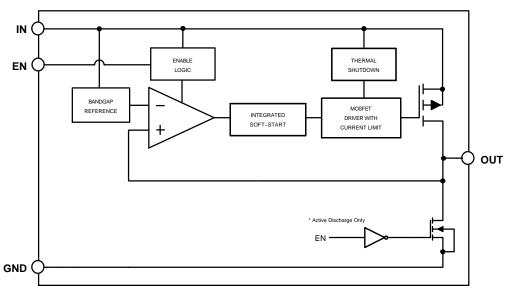


Figure 2. Simplified Schematic Block Diagram

## **PIN FUNCTION DESCRIPTION**

Pin No. CSP4	Pin No. XDFN4	Pin Name	Description
A1	4	IN	Input voltage supply pin
A2	1	OUT	Regulated output voltage. The output should be bypassed with small 1 μF ceramic capacitor.
B1	3	EN	Chip enable: Applying $V_{EN}$ < 0.2 V disables the regulator, Pulling $V_{EN}$ > 0.7 V enables the LDO.
B2	2	GND	Common ground connection
_	EPAD	EPAD	Expose pad can be tied to ground plane for better power dissipation

## **ABSOLUTE MAXIMUM RATINGS**

Rating	Symbol	Value	Unit
Input Voltage (Note 1)	V <sub>IN</sub>	-0.3 V to 6	V
Output Voltage	V <sub>OUT</sub>	-0.3 to V <sub>IN</sub> + 0.3, max. 6 V	V
Chip Enable Input	V <sub>CE</sub>	-0.3 to 6 V	V
Output Short Circuit Duration	t <sub>SC</sub>	unlimited	s
Maximum Junction Temperature	TJ	150	°C
Storage Temperature	T <sub>STG</sub>	-55 to 150	°C
ESD Capability, Human Body Model (Note 2)	ESD <sub>HBM</sub>	2000	V
ESD Capability, Machine Model (Note 2)	ESD <sub>MM</sub>	200	V

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. Refer to ELECTRICAL CHARACTERISTICS and APPLICATION INFORMATION for Safe Operating Area.

- Refer to ELECTRICAL CHARACTERISTICS and APPLICATION INFORMATION for Safe Oper
   This device series incorporates ESD protection and is tested by the following methods:
  - ESD Human Body Model tested per EIA/JESD22–A114
    - ESD Machine Model tested per EIA/JESD22-A115
    - Latchup Current Maximum Rating tested per JEDEC standard: JESD78.

## THERMAL CHARACTERISTICS

Rating		Value	Unit
Thermal Characteristics, CSP4 (Note 3) Thermal Resistance, Junction–to–Air		108	2011
Thermal Characteristics, XDFN4 (Note 3) Thermal Resistance, Junction-to-Air	$R_{ hetaJA}$	208	°C/W

3. Measured according to JEDEC board specification. Detailed description of the board can be found in JESD51-7

**ELECTRICAL CHARACTERISTICS**  $-40^{\circ}\text{C} \le T_J \le 125^{\circ}\text{C}$ ;  $V_{\text{IN}} = V_{\text{OUT(NOM)}} + 0.3 \text{ V or } 1.1 \text{ V, whichever is greater; } I_{\text{OUT}} = 1 \text{ mA, } C_{\text{IN}} = 1 \text{ mA}$  $C_{OUT} = 1 \mu F$ , unless otherwise noted.  $V_{EN} = 1.0 \text{ V}$ . Typical values are at  $T_J = +25 \text{°C}$  (Note 4).

Parameter	Test Conditi	Symbol	Min	Тур	Max	Unit	
Operating Input Voltage			V <sub>IN</sub>	1.1		5.5	V
Output Voltage Accuracy	$V_{IN} = V_{OUT(NOM)} + 0.3 V$ $(V_{IN} \ge 1.1 V)$	V <sub>OUT(NOM)</sub> ≤ 1.5 V	V <sub>OUT</sub>	-30		+30	mV
	(V <sub>IN</sub> ≥ 1.1 V)	V <sub>OUT(NOM) &gt;</sub> 1.5 V		-2		+2	%
Line Regulation	$V_{OUT(NOM)} + 0.5 V \le V_{IN} \le 10$	5.5 V, (V <sub>IN</sub> ≥ 1.1 V)	Line <sub>Reg</sub>		0.02		%/V
Load Regulation	I <sub>OUT</sub> = 1 mA to 2	200 mA	Load <sub>Reg</sub>		0.001		%/mA
Dropout Voltage (Note 5)	V <sub>OUT(NOM)</sub> = 1.05 V	I <sub>OUT</sub> = 50 mA	$V_{DO}$		40	70	mV
		I <sub>OUT</sub> = 100 mA			70	130	1
	V <sub>OUT(NOM)</sub> = 1.20 V	I <sub>OUT</sub> = 110 mA			60	140	1
		I <sub>OUT</sub> = 200 mA			110	190	1
	V <sub>OUT(NOM)</sub> = 1.80 V	I <sub>OUT</sub> = 200 mA			65	120	1
	V <sub>OUT(NOM)</sub> = 2.80 V	I <sub>OUT</sub> = 200 mA			45	100	1
Output Current Limit	V <sub>OUT</sub> = 90% V <sub>OUT(NOM)</sub>		I <sub>CL</sub>	225	300		
Short Circuit Current	V <sub>OUT</sub> = 0	I <sub>SC</sub>		300		mA	
Quiescent Current	I <sub>OUT</sub> = 0 m	IQ		20	25	μΑ	
Shutdown Current	$V_{EN} \le 0.2 \text{ V}, V_{IN} = 1.1 \text{ V}$		I <sub>DIS</sub>		0.01	1.0	μΑ
EN Pin Threshold Voltage	EN Input Volta	V <sub>ENH</sub>	0.7				
	EN Input Volta	V <sub>ENL</sub>			0.2	V	
EN Pull Down Current	V <sub>EN</sub> = 1.1	V	I <sub>EN</sub>		0.2	0.5	μΑ
Turn-On Time	$C_{OUT}$ = 1 $\mu$ F, From asse $V_{OUT}$ = 95% $V_{O}$	ertion of V <sub>EN</sub> to	t <sub>ON</sub>		120		μs
Power Supply Rejection Ratio	I <sub>OUT</sub> = 20 mA, V <sub>IN</sub> = V <sub>OUT</sub> + 0.3 V	f = 100 Hz f = 1 kHz f = 10 kHz f = 100 kHz	PSRR		90 95 85 55		dB
Output Voltage Noise	f = 10 Hz to 100 kHz		V <sub>N</sub>		8.8		$\mu V_{RMS}$
Thermal Shutdown Threshold	Temperature	T <sub>SDH</sub>		160		°C	
	Temperature f	T <sub>SDL</sub>		140		°C	
Active Output Discharge Resistance	V <sub>EN</sub> < 0.2 V, Versi	R <sub>DIS</sub>		280		Ω	

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

<sup>4.</sup> Performance guaranteed over the indicated operating temperature range by design and/or characterization. Production tested at T<sub>A</sub> = 25°C. Low duty cycle pulse techniques are used during the testing to maintain the junction temperature as close to ambient as possible.

Dropout voltage is characterized when V<sub>OUT</sub> falls 0.02 x V<sub>OUT(NOM)</sub> below V<sub>OUT(NOM)</sub>.
 Guaranteed by design.

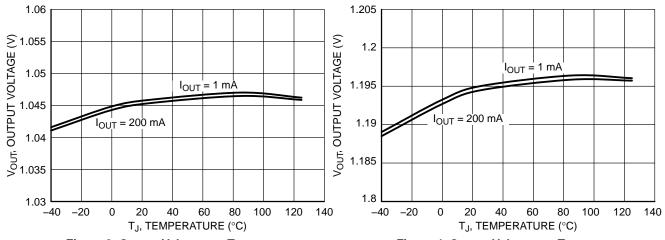


Figure 3. Output Voltage vs. Temperature – V<sub>OUT.nom</sub> = 1.05 V – CSP4

Figure 4. Output Voltage vs. Temperature – V<sub>OUT.nom</sub> = 1.2 V – CSP4

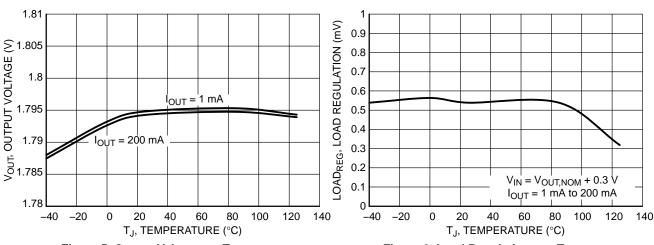


Figure 5. Output Voltage vs. Temperature – V<sub>OUT,nom</sub> = 1.8 V – CSP4

Figure 6. Load Regulation vs. Temperature

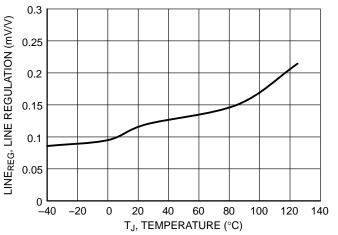


Figure 7. Line Regulation vs. Temperature

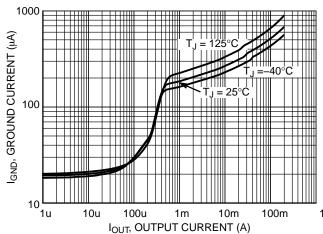


Figure 8. Ground Current vs. Output Current –  $V_{OUT,nom} = 1.2 \text{ V}$ 

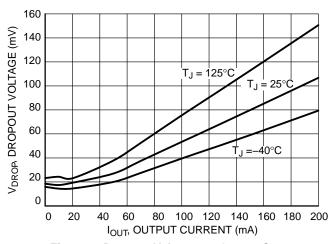


Figure 9. Dropout Voltage vs. Output Current – V<sub>OUT</sub>,nom = 1.2 V – CSP4 Package

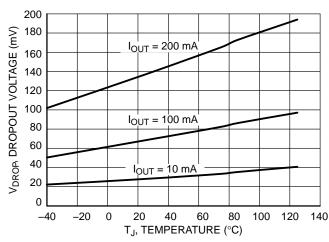


Figure 10. Dropout Voltage vs. Temperature – VOUT, nom = 1.05 V – CSP4 Package

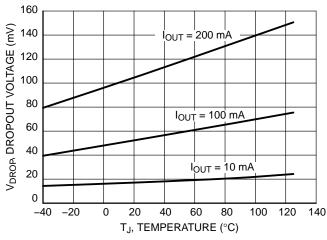


Figure 11. Dropout Voltage vs. Temperature – V<sub>OUT</sub>,nom = 1.2 V – CSP4 Package

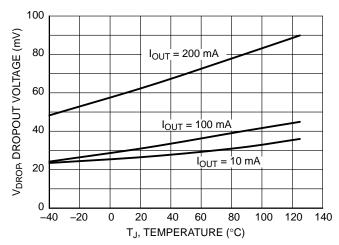


Figure 12. Dropout Voltage vs. Temperature – V<sub>OUT</sub>,nom = 1.8 V – CSP4 Package

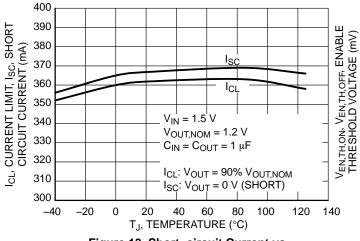


Figure 13. Short-circuit Current vs. Temperature

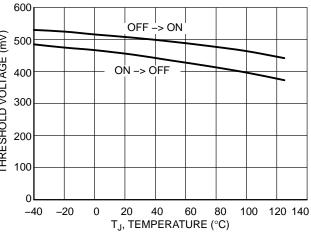
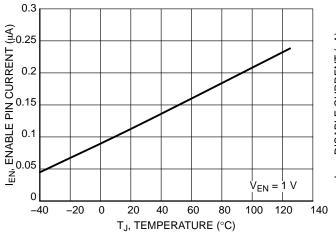


Figure 14. Enable thresholds voltage vs.
Temperature





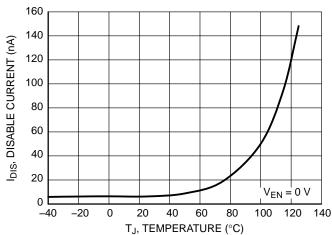


Figure 16. Disable Current vs. Temperature

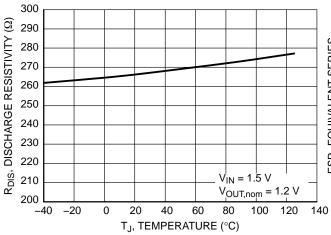


Figure 17. Discharge Resistivity vs.
Temperature

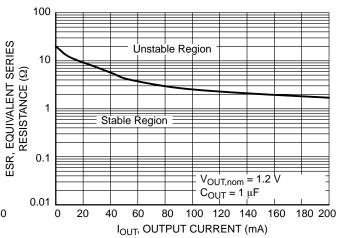
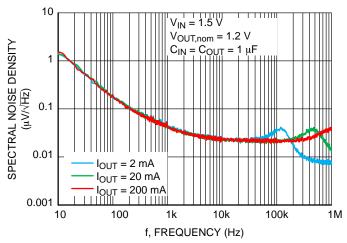
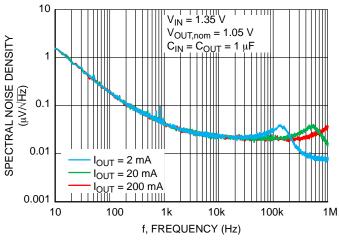


Figure 18. Maximum C<sub>OUT</sub> ESR Value vs. Output Current



RMS Output Noise (μV)			
10 Hz – 100 kHz	100 Hz – 100 kHz		
10.01	8.79		
8.78	7.39		
8.77	7.44		
	10 Hz – 100 kHz 10.01 8.78		

Figure 19. Output Voltage Spectral Noise Density vs. Frequency



 I<sub>OUT</sub> (mA)
 RMS Output Noise (μV)

 10 Hz - 100 kHz
 100 Hz - 100 kHz

 2
 10.01
 8.79

 20
 8.78
 7.39

 200
 8.77
 7.44

Figure 20. Output Voltage Spectral Noise Density vs. Frequency

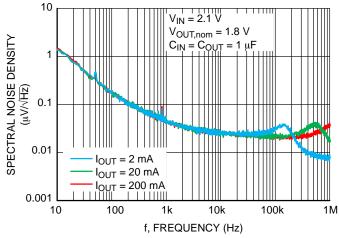


Figure 21. Output Voltage Spectral Noise Density vs. Frequency

I <sub>OUT</sub>	RMS Output Noise (μV)		
(mA)	10 Hz – 100 kHz	100 Hz – 100 kHz	
2	9.88	8.71	
20	9.01	7.73	
200	9.08	7.70	

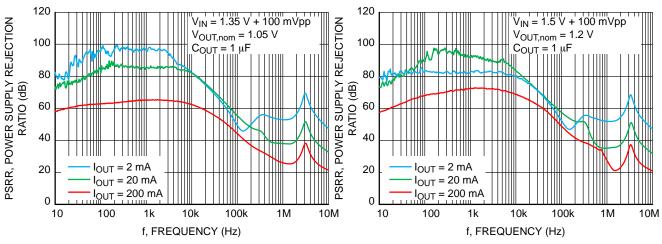


Figure 22. PSRR vs. Frequency

Figure 23. PSRR vs. Frequency

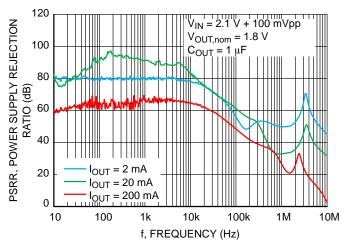


Figure 24. PSRR vs. Frequency

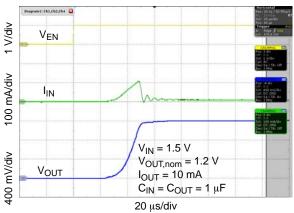
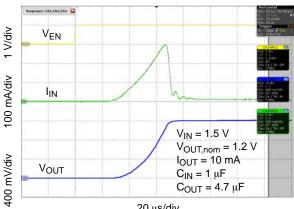


Figure 25. Enable Turn-on Response,  $C_{OUT} = 1~\mu F, \, I_{OUT} = 10~mA$ 



20 μs/div Figure 26. Enable Turn–on Response,  $C_{OUT} = 4.7 \mu F$ ,  $I_{OUT} = 10 mA$ 

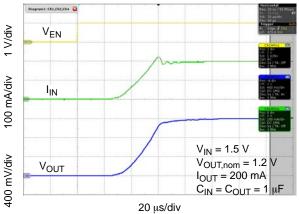


Figure 27. Enable Turn-on Response,  $C_{OUT} = 1 \mu F$ ,  $I_{OUT} = 200 \text{ mA}$ 

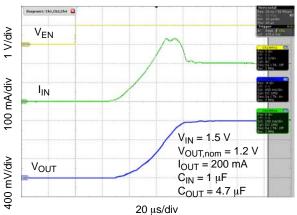


Figure 28. Enable Turn-on Response,  $C_{OUT} = 4.7 \mu F$ ,  $I_{OUT} = 200 \text{ mA}$ 

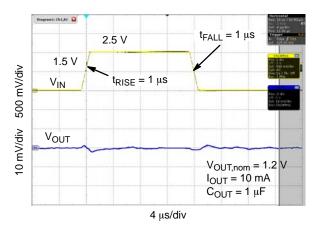


Figure 29. Line Transient Response,  $I_{OUT} = 10 \text{ mA}$ 

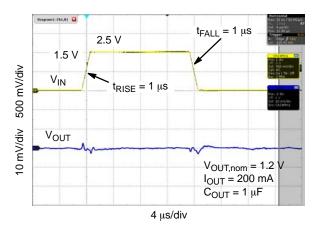


Figure 30. Line Transient Response,  $I_{OUT} = 200 \text{ mA}$ 

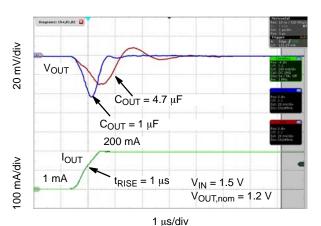


Figure 31. Load Transient Response, I<sub>OUT</sub> = 1 mA to 200 mA

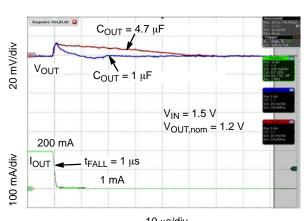


Figure 32. Load Transient Response,  $I_{OUT} = 1$  mA to 200 mA

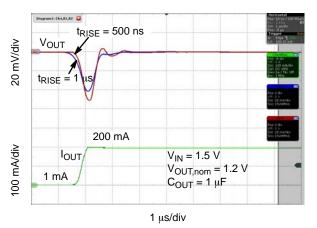


Figure 33. Load Transient Response, I<sub>OUT</sub> = 1 mA to 200 mA

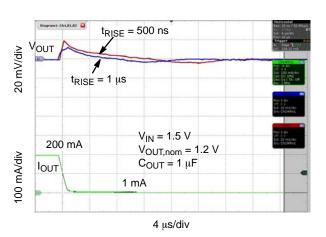


Figure 34. Load Transient Response, I<sub>OUT</sub> = 1 mA to 200 mA

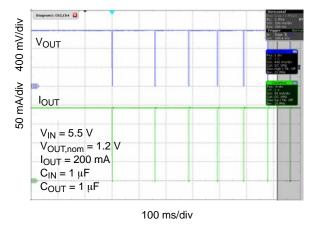


Figure 35. Overheating Protection – TSD

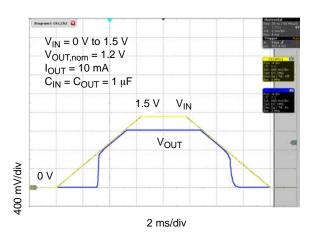


Figure 36. Turn On/Off, Slow Rising V<sub>IN</sub>

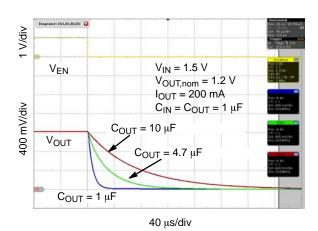


Figure 37. Enable Turn-off Response, Various Output Capacitors

#### APPLICATIONS INFORMATION

#### General

The NCP110 is an ultra-low input voltage, ultra-low noise 200 mA low dropout regulator designed to meet the requirements of low voltage RF applications and high performance analog circuits. The NCP110 device provides very high PSRR and excellent dynamic response. In connection with low quiescent current this device is well suitable for battery powered application such as cell phones, tablets and other. The NCP110 is fully protected in case of current overload, output short circuit and overheating.

## Input Capacitor Selection (CIN)

Input capacitor connected as close as possible is necessary for ensure device stability. The X7R or X5R capacitor should be used for reliable performance over temperature range. The value of the input capacitor should be 1  $\mu F$  or greater to ensure the best dynamic performance. This capacitor will provide a low impedance path for unwanted AC signals or noise modulated onto constant input voltage. There is no requirement for the ESR of the input capacitor but it is recommended to use ceramic capacitors for their low ESR and ESL. A good input capacitor will limit the influence of input trace inductance and source resistance during sudden load current changes.

## **Output decoupling**

The NCP110 requires an output capacitor connected as close as possible to the output pin of the regulator. The recommended capacitor value is  $1\mu F$  and X7R or X5R dielectric due to its low capacitance variations over the specified temperature range. The NCP110 is designed to remain stable with minimum effective capacitance of  $0.6\mu F$  to account for changes with temperature, DC bias and package size. Especially for small package size capacitors such as 0201 the effective capacitance drops rapidly with the applied DC bias. Please refer to Figure 38.

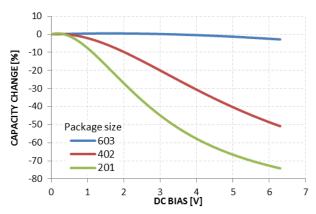


Figure 38. Capacity vs DC Bias Voltage

There is no requirement for the minimum value of Equivalent Series Resistance (ESR) for the  $C_{OUT}$  but the maximum value of ESR should be less than 1.6  $\Omega$ . Larger

output capacitors and lower ESR could improve the load transient response or high frequency PSRR. It is not recommended to use tantalum capacitors on the output due to their large ESR. The equivalent series resistance of tantalum capacitors is also strongly dependent on the temperature, increasing at low temperature.

## **Enable Operation**

The NCP110 uses the EN pin to enable/disable its device and to deactivate/activate the active discharge function. If the EN pin voltage is <0.2 V the device is guaranteed to be disabled. The pass transistor is turned—off so that there is virtually no current flow between the IN and OUT. The active discharge transistor is active so that the output voltage  $V_{OUT}$  is pulled to GND through a 280  $\Omega$  resistor. In the disable state the device consumes as low as typ. 10 nA from the  $V_{IN}$ . If the EN pin voltage >0.7 V the device is guaranteed to be enabled. The NCP110 regulates the output voltage and the active discharge transistor is turned—off. The EN pin has internal pull—down current source with typ. value of 200 nA which assures that the device is turned—off when the EN pin is not connected. In the case where the EN function isn't required the EN should be tied directly to IN.

## **Output Current Limit**

Output Current is internally limited within the IC to a typical 350 mA. The NCP110 will source this amount of current measured with a voltage drops on the 90% of the nominal  $V_{OUT}$ . If the Output Voltage is directly shorted to ground ( $V_{OUT} = 0$  V), the short circuit protection will limit the output current to 360 mA (typ). The current limit and short circuit protection will work properly over whole temperature range and also input voltage range. There is no limitation for the short circuit duration.

## Thermal Shutdown

When the die temperature exceeds the Thermal Shutdown threshold (TSD –  $160^{\circ}$ C typical), Thermal Shutdown event is detected and the device is disabled. The IC will remain in this state until the die temperature decreases below the Thermal Shutdown Reset threshold (TSDU –  $140^{\circ}$ C typical). Once the IC temperature falls below the  $140^{\circ}$ C the LDO is enabled again. The thermal shutdown feature provides the protection from a catastrophic device failure due to accidental overheating. This protection is not intended to be used as a substitute for proper heat sinking.

#### **Power Dissipation**

As power dissipated in the NCP110 increases, it might become necessary to provide some thermal relief. The maximum power dissipation supported by the device is dependent upon board design and layout. Mounting pad configuration on the PCB, the board material, and the ambient temperature affect the rate of junction temperature

rise for the part. The maximum power dissipation the NCP110 can handle is given by:

$$P_{D(MAX)} = \frac{\left[125^{\circ}C - T_{A}\right]}{\theta_{JA}} \tag{eq. 1}$$

The power dissipated by the NCP110 for given application conditions can be calculated from the following equations:

$$P_D \approx V_{IN} \cdot I_{GND} + I_{OUT} (V_{IN} - V_{OUT})$$
 (eq. 2)

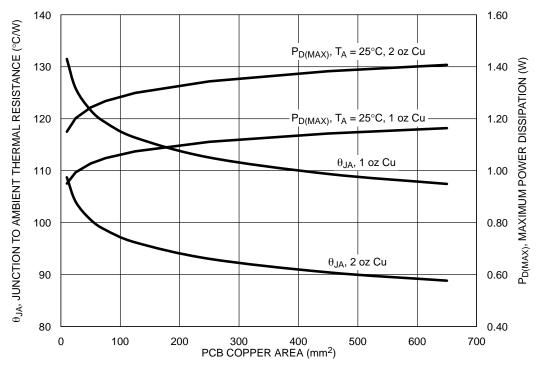


Figure 39.  $\theta_{JA}$  and  $P_{D\,(MAX)}$  vs. Copper Area (CSP4)

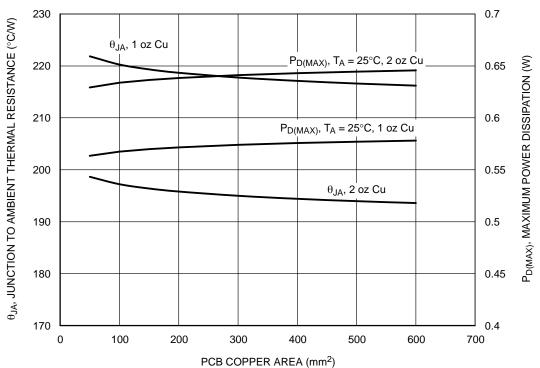


Figure 40.  $\theta_{JA}$  and  $P_{D\;(MAX)}$  vs. Copper Area (XDFN4)

## **ORDERING INFORMATION**

	Nominal Output					
Device	Voltage	Marking	Rotation	Description	Package	Shipping <sup>†</sup>
NCP110AFCT060T2G	0.60 V	С	0°			
NCP110AFCT080T2G	0.80 V	J	0°			
NCP110AFCT085T2G	0.85 V	2	0°			
NCP110AFCT100T2G	1.00 V	Т	0°		WLCSP4	5000 /
NCP110AFCT105T2G	1.05 V	Α	0°	200 mA, Active Discharge	CASE 567VS Tap	Tape & Reel
NCP110AFCT110T2G	1.10 V	G	0°			
NCP110AFCT120T2G	1.20 V	F	0°			
NCP110AFCT180T2G	1.80 V	D	0°			
NCP110AFCT280T2G	2.80 V	E	0°			

## **ORDERING INFORMATION**

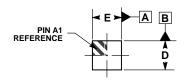
Device	Nominal Output Voltage	Marking	Description	Package	Shipping
NCP110AMX060TBG	0.60 V	FC			
NCP110AMX075TBG	0.75 V	F3			3000 / Tape & Reel
NCP110AMX080TBG	0.80 V	FJ			
NCP110AMX085TBG	0.85 V	F2		XDFN4 CASE 711AJ (Pb-Free)	
NCP110AMX100TBG	1.00 V	FG	200 mA, Active Discharge		
NCP110AMX105TBG	1.05 V	FA			
NCP110AMX110TBG	1.10 V	FH			
NCP110AMX120TBG	1.20 V	FF			
NCP110AMX180TBG	1.80 V	FD			
NCP110AMX280TBG	2.80 V	FE			

<sup>†</sup>For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

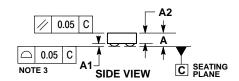
## **PACKAGE DIMENSIONS**

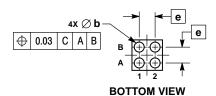
## WLCSP4, 0.64x0.64x0.33

CASE 567VS ISSUE O



**TOP VIEW** 

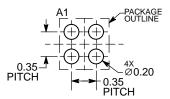




- NOTES:
  1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
  2. CONTROLLING DIMENSION: MILLIMETERS.
  3. COPLANARITY APPLIES TO SPHERICAL CROWNS OF SOLDER BALLS.

	MILLIMETERS				
DIM	MIN NOM MAX				
Α	0.33				
A1	0.04 0.06 0.08				
A2		0.23 REF			
b	0.180	0.200	0.220		
D	0.610	0.640	0.670		
E	0.610	0.640	0.670		
е	0.35 BSC				

# RECOMMENDED SOLDERING FOOTPRINT\*

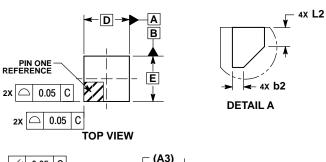


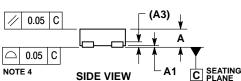
DIMENSIONS: MILLIMETERS

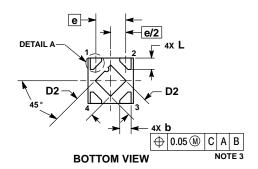
\*For additional information on our Pb–Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

## PACKAGE DIMENSIONS

## XDFN4 1.0x1.0, 0.65P CASE 711AJ **ISSUE A**





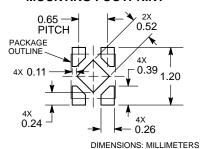


#### NOTES

- DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994. CONTROLLING DIMENSION: MILLIMETERS. DIMENSION 6 APPLIES TO PLATED TERMINAL AND IS MEASURED BETWEEN 0.15 AND
- 0.20 mm FROM THE TERMINAL TIPS. COPLANARITY APPLIES TO THE EXPOSED PAD AS WELL AS THE TERMINALS.

	MILLIM	MILLIMETERS			
DIM	MIN	MAX			
Α	0.33	0.43			
A1	0.00	0.05			
A3	0.10	REF			
b	0.15	0.25			
b2	0.02	0.12			
D	1.00	BSC			
D2	0.43	0.53			
E	1.00	BSC			
е	0.65 BSC				
L	0.20	0.30			
L2	0.07	0.17			

## **RECOMMENDED MOUNTING FOOTPRINT\***



\*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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