



#### 8-channel PCI Express Gen 3, 10GbE, SATA3 ReDriver™ with Linear Equalization

#### **Features**

- → High-speed serial link with linear equalizer
- → Support PCIe Gen 1/2/3 protocol, 10GbE, SATA3, SAS3
- → Supporting 8 differential channels
- → Independent channel configuration of receiver equalization, output swing and flat gain
- → Per Channel Activity Detector with selectable input termination between  $50\Omega$  to  $V_{DD}$  and  $200K\Omega$  to  $V_{DD}$
- $\Rightarrow$  Per Channel Output Termination Detector on power up with selectable output termination between 50 $\Omega$  to V<sub>DD</sub> and High impedance
- → Very linear transfer function
- → Fully compliant to PCISIG Link Training
- → Single-ended mode receiver detection for PCIe
- → Input Threshold detection
- → Pin strap and I<sup>2</sup>C master/slave selectable device programming with external EEPROM
- → 4-bit selectable address bit for I<sup>2</sup>C
- → Supply Voltage: 3.3V±0.3V
- → Industrial Temperature Range: -40°C to 85°C
- → Packaging (Pb-free & Green):
  - 54-contact TQFN (10mm x 5.5mm x 0.5mm pitch) flowthrough pinout

#### **Description**

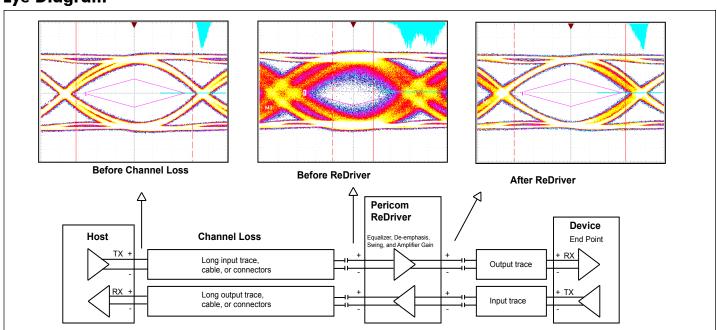
The PI3EQX12908A2 is an 8-channel PCIe Gen 2, 10GbE, SATA3, ReDriver. The device provides programmable linear equalization, output swing and gain, by either pin strapping option or  $I^2C$  Control, to optimize performance over a variety of physical mediums by reducing Inter-symbol interference.

The PI3EQX12908A2 supports eight 100-Ohm Differential CML data I/O's and extends the signals across other distant data pathways on the user's platform.

The programmable settings can be applied easily via pins, software (I<sup>2</sup>C) or loaded via an external EEPROM. When operating in the EEPROM mode, the configuration information is automatically loaded on power up, which eliminates the need for an external microprocessor or software driver.

The PI3EQX12908A2 offers fully Linear Transfer function to fully comply with all PCIe 3 Link Training signals.

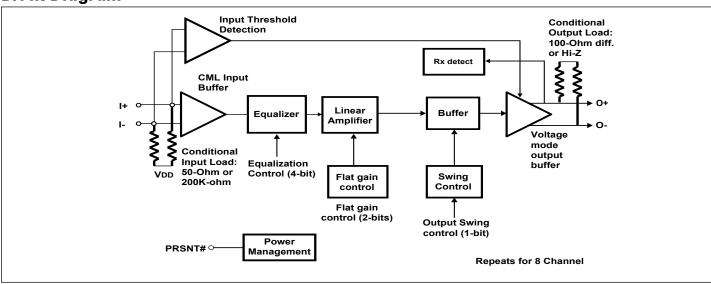
## **Eye Diagram**



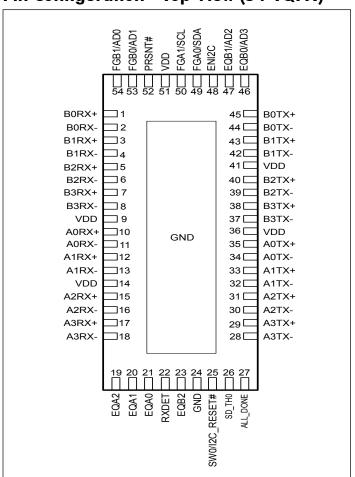




## **Block Diagram**



## Pin Configuration - Top View (54-TQFN)







# **Pin Description**

(Flow-Thru Pinout)

Pin # (54-TQFN)	Pin Name	Type	Description	
Data Signals				
10	A0RX+	I	Differential inputs for Channel A0, with internal 50-Ohm pull-up and >200K-Ohm	
11	A0RX-	I	otherwise.	
35	A0TX+,	О		
34	A0TX-	О	Differential outputs for Channel A0	
12	A1RX+,	I	Differential inputs for Channel A1, with internal 50-Ohm pull-up and >200K-Ohm	
13	A1RX-	I	otherwise.	
33	A1TX+,	О	Differential outputs for Channel A1	
32	A1TX-	О	Differential outputs for Channel A1	
15	A2RX+,	I	Differential inputs for Channel A2, with internal 50-Ohm pull-up and >200K-Ohm	
16	A2RX-	I	otherwise.	
31	A2TX+,	О	Differential continues for Channel A2	
30	A2TX-	О	Differential outputs for Channel A2	
17	A3RX+,	I	Differential inputs for Channel A3, with internal 50-Ohm pull-up and >200K-Ohm	
18	A3RX-	I	otherwise.	
29	A3TX+,	О	Differential outputs for Channel A2	
28	A3TX-	О	Differential outputs for Channel A3	
1	B0RX+,	I	Differential inputs for Channel B0, with internal 50-Ohm pullup and >200KOhm	
2	B0RX-	I	otherwise.	
45	B0TX+,	О	Differential outputs for Channel B0	
44	B0TX-	О	Differential outputs for Chairner Bo	
3	B1RX+,	I	Differential inputs for Channel B1, with internal 50-Ohm pullup and >200KOhm	
4	B1RX-	I	otherwise.	
43	B1TX+,	О	Differential outputs for Channel B1	
42	B1TX-	О	Differential outputs for Chairner B1	
5	B2RX+,	I	Differential inputs for Channel B2, with internal 50-Ohm pullup and >200KOhm	
6	B2RX-	I	otherwise.	
40	B2TX+,	О	Differential outputs for Channel P2	
39	B2TX-	О	Differential outputs for Channel B2	
7	B3RX+,	I	Differential inputs for Channel B3, with internal 50-Ohm pullup and >200KOhm	
8	B3RX-	I	otherwise.	
38	B3TX+,	О	Differential outputs for Channel B3	
37	B3TX-	О	Differential outputs for Chairner by	





Pin Description Cont.

Pin # (54-TQFN)	Pin Name	Type	Description
Control Signals		-/1-	
48	ENI2C	I	$\rm I^2C$ Enable Pin. When tied to Low, each channel is programmed by the external pin voltage (Pin Mode). When tied to High, each channel is programmed by the data stored in the $\rm I^2C$ bus (Slave Mode). When FLOAT, data is accesses from external EEPROM (Master Mode). ENI2C has pull-up / pull-down 90k-Ohm resistance (Default = $\rm V_{DD}$ / 2).
When $ENI2C = 1$ (I	<sup>2</sup> C Mode)		
50	SCL	I/O	$I^2C$ SCL clock input in $I^2C$ Slave Mode (ENI2C = High). This pin becomes clock output when loading from EEPROM in $I^2C$ Master Mode (ENI2C = FLOAT).
49	SDA	I/O	I <sup>2</sup> C SDA data input/output in I <sup>2</sup> C Master or Slave Mode.
54, 53, 47, 46	AD[0:3]	I	$I^2C \ programmable \ address \ bits \ in \ I^2C \ Master \ or \ Slave \ Mode. \ AD[0:2] \ have \ pullup 90k-Ohm \ resistance.$ $(Default = V_{DD}/2)$
25	I2C_RESET#	I	Reset Pin for $I^2C$ . When Low, the registers are reset to default value. $I^2C$ _RESET# has pull-up 90k-Ohm resistance
When $ENI2C = 0$ (I	Pin mode)		
21, 20, 19	EQA[0:2]	I	These pins set the level of Equalizer in Bank A channels when ENI2C is Low. When ENI2C is High, the I2C registers provide independent control of each channel. See Table 1: Equalizer Settings. EQA[1] has pull-up 90k-Ohm resistance. EQA[0] and EQA[2] have pull-up / pull-down 90k-Ohm resistance (Default = $V_{\rm DD}$ / 2).
46, 47, 23	EQB[0:2]	I	These pins set the level of Equalizer in Bank B channels when ENI2C is Low. When ENI2C is High, the I2C registers provide independent control of each channel, and the EQB[1:0] pins are converted to I <sup>2</sup> C AD[2:3] inputs. See Table 1: Equalizer Settings. EQB[1] has pull-up 90k-Ohm resistance. EQB[0] and EQB[2] have pull-up / pull-down 90k-Ohm resistance (Default = $V_{\rm DD}$ / 2).
49, 50	FGA[0:1]	I	These pins control the level of Flat Gain in Bank A channels when ENI2C is Low. When ENI2C is High, the I <sup>2</sup> C registers provide independent control of each channel, and the FGA[1:0] pins are converted to I <sup>2</sup> C SCL/SDA. See Table 2: Flat Gain Settings.
53, 54	FGB[0:1]	I	These pins control the level of Flat Gain in Bank B channels when ENI2C is Low. When ENI2C is High, the I <sup>2</sup> C registers provide independent control of each channel, and FGB[1:0] pins are converted to AD[0:1] inputs. See Table 2: Flat Gain Settings. FGB[0] and FGB[1] have pull-up 90k-Ohm resistance.
25	SW0	I	This pins sets the Output Voltage Swing level in all channels when $ENI^2C$ is Low. SW0 has pull-up 90k-Ohm resistance.
26	SD_TH0	I	Internal Signal Detect Threshold. This pin should be tied to $V_{\rm DD}$ for normal operation. Refer to Table 4 for more options. SD_TH0 has pull-up 90k-Ohm resistance.
In both I2C and Pir	n modes		
22	RXDET	I	Receiver Detection Control Pin. When High, receiver detection is enabled to support PCIe Mode. When Low, receiver detection is disabled to support 10GbE and SATA3 Modes and input is 50-Ohm to $V_{\rm DD}$ . RXDET has pull-up 90k-Ohm resistance.





# Pin Description Cont.

Pin # (54-TQFN)	Pin Name	Type	Description			
52	PRSNT#	I	Cable Present Detect Input. When High, a cable is not present per PCIe Cabling Specification 1.0, and the device is put in lower power mode. When Low, the device is enabled and in normal operation. PRSNT# has pull-up 90k-Ohm resistance.			
Output	Output					
27	ALL_DONE	О	Valid Register Load Status Output. When LOW, the external EEPROM load has failed. When HIGH, the external EEPROM load is successful.			
Power Pins						
9, 14, 36, 41, 51	$V_{\mathrm{DD}}$	PWR	3.3V ± 10% Supply Voltage			
Center Pad, 24	GND	PWR	Supply GND			

5





## **Description of Operation**

#### **Output Receiver Detector:**

On power up or when PRSNT# becomes low, the output resistance is set to high impedance, and the input resistance is set to 200K ohms. The device continually looks to detect an external 50 ohm termination resistor on a per channel basis. If no 50 ohms is detected in the first 40us of time, the channel is continually polled with 40us detection cycle until detection occurs. This operation can only be reinitiated when PRSNT# or I2C\_RESET# are toggled again.

#### **Input Activity Detector:**

When the input voltage on individual channel basis falls below Vth-, the output is driven to the common mode voltage so as to eliminate output chatter. When the input voltage is higher than Vth+, the channel is resumed immediately.

#### **Power Enable function:**

One pin control or I2C control, when PRSNT# is set to high, the IC goes into power down mode, both input and output termination set to 200K and high impedance respectively. Individual channel enabling is done through the I2C register programming.

### **Equalization Setting:**

EQA[2:0] and EQB[2:0] are the selection pins for the equalization selection for each of the channels of A and B respectively.

**Table 1. Equalization Setting** 

	Equalizer setting (dB)									
EQ2	EQ1	EQ0	EQ 4 Bits	@ 1.25GHz	@ 2.5GHz	@ 3GHz	@ 4GHz	@ 5GHz	@ 6GHz	
0	0	0	0000	0	0.6	1.0	1.5	2.4	2.8	
0	0	1	0001	0.15	1.2	1.7	2.4	3.5	4	
0	1	0	0010	0.4	1.9	2.5	3.3	4.5	5	
0	1	1	0011	0.6	2.5	3.2	4.2	5.3	5.8	
1	0	0	0100	1.8	3.4	4.1	4.9	6.0	6.4	
1	0	1	0101	2.1	3.9	4.7	5.6	6.7	7.1	
1	1	0	0110	2.3	4.4	5.2	6.2	7.3	7.7	
1	1	1	0111	2.5	4.9	5.7	6.8	7.9	8.2	
HIZ	0	0	1000	3.4	5.6	6.4	7.3	8.4	8.7	
HIZ	0	1	1001	3.6	6.0	6.9	7.8	8.9	9.1	
HIZ	1	0	1010	3.8	6.4	7.3	8.3	9.3	9.5	
HIZ	1	1	1011	4.1	6.8	7.7	8.7	9.7	9.9	
0	0	HIZ	1100	5.1	7.5	8.3	9.2	10.1	10.2	
0	1	HIZ	1101	5.3	7.8	8.6	9.5	10.4	10.5	
1	0	HIZ	1110	5.4	8.1	8.9	9.8	10.7	10.8	
1	1	HIZ	1111	5.6	8.4	9.2	10.1	11	11.1	





## Flat Gain Setting:

Flat Gain settings: FGA[0:1] and FGB[0:1] are the selection bits for Flat Gain value for A and B channels.

Table 2. Flat Gain Setting

FGA1 FGB1	FGA0 FGB0	(dB)
0	0	-4
0	1	-2
1	0	0
1	1	2

## **Output Swing Setting:**

SW0 is the selection bit for output swing for A and B channels.

**Table 3. Output Swing Setting** 

SW0	mVp-p	
0	900	
1	1,000	

# **Signal Detect Threshold Level:**

# Table 4. Signal Detect Threshold Level Setting via I<sup>2</sup>C Bus Mode

SD_TH1 <i<sup>2C bit&gt;</i<sup>	SD_TH0	Threshold ON (mVppd)	Threshold OFF (mVppd)
0	0	130	30
0	1	150	50
1	0	170	70
1	1	210	110





# I<sup>2</sup>C Programming

Address assignment							
A6	A5	A4	A3	A2	A1	A0	R/W
1	1	1	AD3	AD2	AD1	AD0	1=R, 0=W

BYTE 0				
Bit	Type	Power up condition	Control affected	Comment
7	R		A3 Signal Detector Output	
6	R		A2 Signal Detector Output	
5	R		A1 Signal Detector Output	
4	R		A0 Signal Detector Output	1= Activity 0=no
3	R		B3 Signal Detector Output	activity
2	R		B2 Signal Detector Output	
1	R		B1 Signal Detector Output	
0	R		B0 Signal Detector Output	
BYTE 1				
Bit	Type	Power up condition	Control affected	Comment
7	R		A3 RX Detector Output	
6	R		A2 RX Detector Output	
5	R		A1 RX Detector Output	
4	R		A0 RX Detector Output	1 = Far-end 50 -ohm detected
3	R		B3 RX Detector Output	0 = Not detected
2	R		B2 RX Detector Output	0 = Not detected
1	R		B1 RX Detector Output	
0	R		B0 RX Detector Output	
BYTE 2				
Bit	Туре	Power up condition	Control affected	Comment
7	R/W	0	A3 Power down	
6	R/W	0	A2 Power down	
5	R/W	0	A1 Power down	

4

3

2

1

0

R/W

R/W

R/W

R/W

R/W

0

0

0

0

0

1 = Power down

A0 Power down

B3 Power down

B2 Power down

B1 Power down

B0 Power down





# I<sup>2</sup>C Programming Cont.

BYTE 3					
Bit	Type	Power up condition		Control affected	Comment
7	R/W	0		EQ3	
6	R/W	0		EQ2	Equalizer
5	R/W	0		EQ1	See Table 1
4	R/W	0	Channel A0 configuration	EQ0	
3	R/W	0		FG1	Flat Gain
2	R/W	0		FG0	See Table 2
1	R/W	0		Reserved	Swing
0	R/W	0		SW0	See Table 3
BYTE 4					
Bit	Type	Power up condition	1	Control affected	Comment
7	R/W	0		EQ3	
6	R/W	0		EQ2	Equalizer
5	R/W	0		EQ1	See Table 1
4	R/W	0	Channel A1 con-	EQ0	
3	R/W	0	figuration	FG1	Flat Gain
2	R/W	0		FG0	See Table 2
1	R/W	0		Reserved	Swing
0	R/W	0		SW0	See Table 3
BYTE 5					
Bit	Type	Power up condition		Control affected	Comment
7	R/W	0		EQ3	
6	R/W	0	Channel A2 con-	EQ2	Equalizer
5	R/W	0	figuration	EQ1	See Table 1
4	R/W	0		EQ0	
3	R/W	0		FG1	Flat Gain
2	R/W	0		FG0	See Table 2
1	R/W	0		Reserved	Swing
0	R/W	0		SW0	See Table 3





# I<sup>2</sup>C Programming Cont.

BYTE 6	5				
Bit	Type	Power up condition		Control affected	Comment
7	R/W	0		EQ3	
6	R/W	0		EQ2	Equalizer
5	R/W	0	Channel A3 con-	EQ1	See Table 1
4	R/W	0 Cł		EQ0	
3	R/W	0 fig	guration	FG1	Flat Gain
2	R/W	0		FG0	See Table 2
1	R/W	0		Reserved	Swing
0	R/W	0		SW0	See Table 3
BYTE 7	7				·
Bit	Type	Power up condition		Control affected	Comment
7	R/W	0	Channel B0 configuration	EQ3	
6	R/W	0		EQ2	Equalizer
5	R/W	0		EQ1	See Table 1
4	R/W	0 Cł		EQ0	
3	R/W	0 fig		FG1	Flat Gain
2	R/W	0		FG0	See Table 2
1	R/W	0		Reserved	Swing
0	R/W	0		SW0	See Table 3
BYTE 8	3				
Bit	Type	Power up condition		Control affected	Comment
7	R/W	0		EQ3	
6	R/W	0		EQ2	Equalizer
5	R/W	0		EQ1	See Table 1
4	R/W	0 Ch	Channel B1 configuration	EQ0	
3	R/W	0 fig		FG1	Flat Gain
2	R/W	0		FG0	See Table 2
1	R/W	0		Reserved	Swing
0	R/W	0		SW0	See Table 3





# I<sup>2</sup>C Programming Cont.

BYTE 9	)				
Bit	Type	Power up condition		Control affected	Comment
7	R/W	0		EQ3	
6	R/W	0		EQ2	Equalizer
5	R/W	0		EQ1	See Table 1
4	R/W	0 Cl	Channel B2 configuration	EQ0	
3	R/W	0 fig		FG1	Flat Gain
2	R/W	0		FG0	See Table 2
1	R/W	0		Reserved	Swing
0	R/W	0		SW0	See Table 3
BYTE 1	10				
Bit	Type	Power up condition		Control affected	Comment
7	R/W	0		EQ3	
6	R/W	0		EQ2	Equalizer
5	R/W	0		EQ1	See Table 1
4	R/W	0 Cl	hannel B3 con-	EQ0	
3	R/W	0 fig	figuration	FG1	Flat Gain
2	R/W	0		FG0	See Table 2
1	R/W	0		Reserved	Swing
0	R/W	0		SW0	See Table 3
BYTE 1	1				
Bit	Type	Power up condition		Control affected	Comment
7	R/W	0		A3 Signal Detector	
6	R/W	0		A2 Signal Detector	
5	R/W	0		A1 Signal Detector	
4	R/W	0		A0 Signal Detector	1=Power Down
3	R/W	0		B3 Signal Detector	1-Fower Down
2	R/W	0		B2 Signal Detector	
1	R/W	0		B1 Signal Detector	
0	R/W	0		B0 Signal Detector	





# I<sup>2</sup>C Programming Cont.

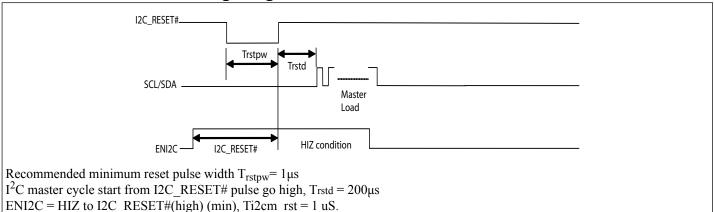
DITE	2			
Bit	Type	Power up condition	Control affected	Comment
7	R/W	0	A3 RX Detector	
6	R/W	0	A2 RX Detector	
5	R/W	0	A1 RX Detector	1=Power Down
4	R/W	0	A0 RX Detector	
3	R/W	0	B3 RX Detector	
2	R/W	0	B2 RX Detector	
1	R/W	0	B1 RX Detector	1=disable
0	R/W	0	B0 RX Detector	
			<del></del>	
BYTE 1	3			
	3 Type	Power up condition	Control affected	Comment
Bit		Power up condition	Control affected	Comment
<b>Bit</b> 7	Туре		Control affected	Comment
<b>BYTE 1 Bit</b> 7 6 5	Type R/W	0		Comment
<b>Bit</b> 7	Type R/W R/W	0 0	Control affected  Reserved	Comment
<b>Bit</b> 7 6 5	Type R/W R/W R/W	0 0 0		Comment
<b>Bit</b> 7 6 5 4	Type R/W R/W R/W R/W	0 0 0 0		Comment
Bit 7 6 5 4 3	Type R/W R/W R/W R/W R/W	0 0 0 0 0		Comment  Signal Detector

12





# Reset and I<sup>2</sup>C Master Timing Diagram



# I<sup>2</sup>C Operation

The integrated I2C interface operates as a master or slave device depending on the pin ENI2C being HIZ or HIGH respectively. Standard mode (100Kbps) is supported with 7-bit addressing. The data byte format is 8-bit bytes, and supports the format of indexing to be compatible with other bus devices. In the Slave mode (ENI2C = HIGH), the device supports Read/Write. The bytes must be accessed in sequential order from the lowest to the highest byte with the ability to stop after any complete byte has been transferred. Address bits A3 to A0 are programmable to support multiple chips environment. The Data is loaded until a Stop sequence is issued.

In the master mode (ENI2C = HIZ), PI3EQX12908A2 supports up to 16 masters connected in daisy chain through connecting I2C\_DONE pin to I2C\_RESET# pin of the next part.

Master EEPROM data starting address of the device address is indicated in the table below:

AD3, AD2, AD1, AD0	<b>EEPROM Data Starting Location</b>
0000	00h
0001	10h
0010	20h
0011	30h
0100	40h
0101	50h
0110	60h
0111	70h
1000	80h
1001	90h
1010	A0h
1011	B0h
1100	C0h
1101	D0h
1110	E0h
1111	F0h





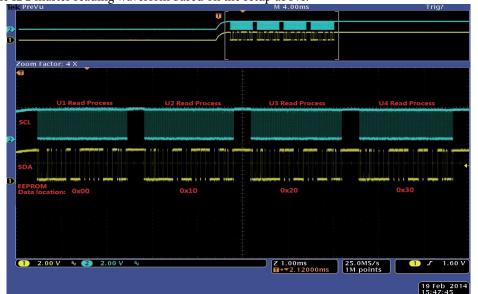
When tying multiple PI3EQX12908A2 devices to the SDA and SCL bus, use the guidelines below to configure the devices. The user also can refer the application notes for detail information.

- Use AD[3:0] address bits so that each device can loaded it's configuration from the EEPROM. Example below is for 4 devices. The first device in the sequence must be address 0x00h; subsequent devices must follow the address order listed below.
- U1: AD[3:0] = 0000 = 0x00h,
- U2: AD[3:0] = 0001 = 0x10h,
- U3: AD[3:0] = 0010 = 0x20h,
- U4: AD[3:0] = 0011 = 0x30h
- For I2C Slave Mode operation, use a 2Kohms pull-up resistor on SDA and SCL pins. For I2C Master Mode operation, use a 1Kohm pull-up resistor on SDA and SCL pins.
- Daisy-chain I2C\_RESET# and ALL\_DONE from one device to the next device in the sequence so that they do not compete for the EEPROM at the same time.
- 1. Tie ALL DONE of U1 to I2C RESET# of U2
- 2. Tie ALL\_DONE of U2 to I2C\_RESET# of U3
- 3. Tie ALL DONE of U3 to I2C RESET# of U4
- 4. Optional: Tie ALL\_DONE output of U4 to a LED to show the devices has been loaded successfully Below is an example of a 2 kbits (256 x 8-bit) EEPROM in hex format for 4pcs PI3EQX12908A2 device. Bold fonts in yellow are register setting from Byte0 to Byte15 for each device in each line. Bold fonts in red is the EEPROM data location.

```
:1000<mark>00</mark>00<mark>000000FF0000FFFFFFFFFF0000FF6201</mark>94
:10001000000000FFFF0000FFFFFFF0000FF620184
:10002000<mark>000000FFFFFF0000FFFFFF0000FF6201</mark>74
:1000<mark>30</mark>00<mark>000000FFFFFFFF0000FFFF0000FF6201</mark>64
:0000001FF
```

2k bits (256 x 8-bit) EEPROM Date Example

Below is the sample of the I2C master reading waveform based on the setup above.



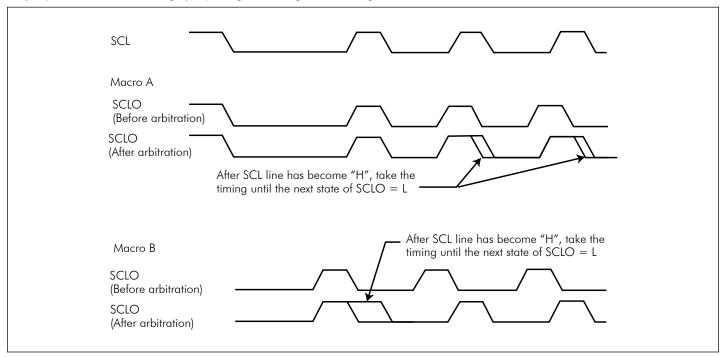
14





#### **SCL Synchronization**

When more than one I2C device becomes a master device and drives the SCL line, each device senses the state of SCL line and automatically adjust the the drive timing by adjusting the timing to the slowest one.



## **Transferring Data**

Every byte put on the SDA line must be 8-bits long. Each byte has to be followed by an acknowledge bit. Data is transferred with the most significant bit (MSB) first (see the I2C Data Transfer diagram). The PI3EQX12908A2 will never hold the clock line SCL LOW to force the master into a wait state.

## **Acknowledge**

Data transfer with acknowledge is required from the master. When the master releases the SDA line (HIGH) during the acknowledge clock pulse, the PI3EQX12908A2 will pull down the SDA line during the acknowledge clock pulse so that it remains stable LOW during the HIGH period of this clock pulse as indicated in the I2C Data Transfer diagram. The PI3EQX12908A2 will generate an acknowledge after each byte has been received.

#### **Data Transfer**

A data transfer cycle begins with the master issuing a start bit. After recognizing a start bit, the PI3EQX12908A2 will watch the next byte of information for a match with its address setting. When a match is found it will respond with a read or write of data on the following clocks. Each byte must be followed by an acknowledge bit, except for the last byte of a read cycle which ends with a stop bit. For a write cycle, the first data byte following the address byte is an index byte that is used by the PI3EQX12908A2. Data is transferred with the most significant bit (MSB) first.

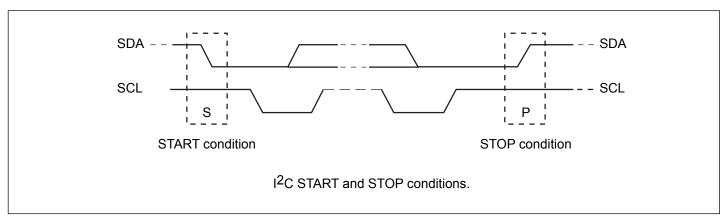




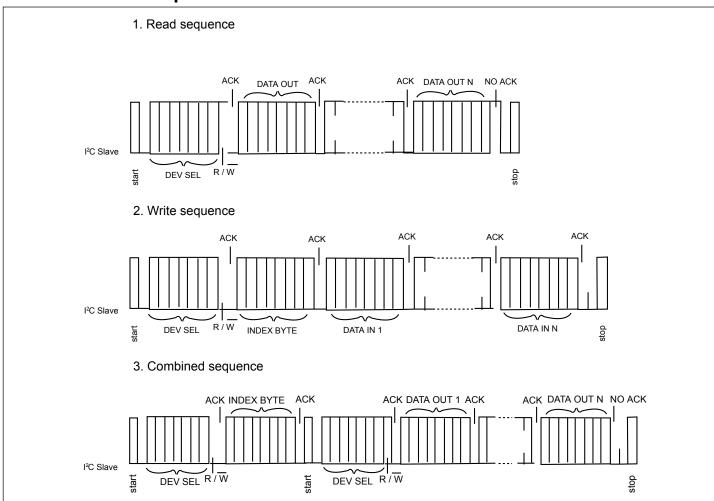
# I<sup>2</sup>C Data Transfer

## **Start & Stop Conditions**

A HIGH to LOW transition on the SDA line while SCL is HIGH indicates a START condition. A LOW to HIGH transition on the SDA line while SCL is HIGH defines a STOP condition, as shown in the figure below. When a STOP condition is detected, index byte value will be reset to 0.



# I<sup>2</sup>C Data Transfer Sequence



16





# **Maximum Ratings**

(Above which useful life may be impaired. For user guidelines, not tested.)

· -	
Supply Voltage	0.5V to +4.0V
LVCMOS Input/Output Voltage	0.5V to +4.0V
CML Input Voltage	0.5V to (V <sub>DD</sub> +0.5)
CML Input Current	30 to +30 mA
I <sup>2</sup> C pins	V <sub>DD</sub> +0.3 V
Storage Temperature	65°C to +150°C
Max. Junction Temperature	125 °C
ESD HBM	2kV

#### Note:

Stresses greater than those listed under MAXIMUM RATINGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

## **LVCMOS I/O DC Specifications** ( $V_{DD} = 3.3 \pm 10\%$ , $T_A = -40 \text{ to } 85^{\circ}\text{C}$ )

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
$V_{\mathrm{IH}}$	DC input logic high		V <sub>DD</sub> /2 + 0.7		$V_{\mathrm{DD}} + 0.3$	V
V <sub>IL</sub>	DC input logic low		-0.3		V <sub>DD</sub> /2 - 0.7	V
V <sub>OH</sub>	DC output logic high	At $I_{OH} = -200 \mu A$	V <sub>DD</sub> - 0.2			V
V <sub>OL</sub>	DC output logic low	At $I_{OL} = +200 \mu A$			0.2	V
V <sub>hys</sub>	Hysteresis of Schmitt trigger input		0.8			V

# SDA and SCL I/O for I<sup>2</sup>C-bus ( $V_{DD}=3.3\pm10\%$ , $T_{A}=-40$ to $85^{\circ}C$ )

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
V <sub>IH</sub>	DC input logic high		V <sub>DD</sub> /2 + 0.7		$V_{\rm DD} + 0.3$	V
$V_{\rm IL}$	DC input logic low		-0.3		V <sub>DD</sub> /2 - 0.7	V
V <sub>OL</sub>	DC output logic low	$I_{OL} = 3mA$			0.4	V
V <sub>hys</sub>	Hysteresis of Schmitt trigger input		0.8			V
t <sub>of</sub>	Output fall time from $V_{IHmin}$ to $V_{ILmax}$	Bus capacity = 10 to 400pF		250		ns
f <sub>SCLK</sub>	SCLK clock frequency			100		kHz

# High Speed I/O AC/DC Specifications ( $V_{DD}=3.3\pm10\%$ , $T_A=-40$ to $85^{\circ}$ C)

Receiver Inp	Receiver Input (100 $\Omega$ differential)					
Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
C <sub>RX</sub>	RX AC coupling capacitance			220		nF
S <sub>11</sub>	Innut natura loca	10MHz to 4GHz dif- ferential		13		ηn
	Input return loss	1GHz to 4GHz common mode		4		dB





# High Speed I/O AC/DC Specifications Cont.

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units	
C	Output voture loss	10MHz to 4GHz differential		21		JD.	
S <sub>22</sub>	Output return loss	1GHz to 4GHz common mode		4		dB	
D	DC single-ended input impedance			50		0	
$R_{IN}$	DC differential input impedance			100		Ω	
D	DC single-ended output impedance			50		0	
$R_{OUT}$	DC differential output impedance			100		Ω	
Z <sub>RX-HIZ</sub>	DC input CM input impedance during reset or power down			200		kΩ	
V <sub>RX-DIFF-PP</sub>	Differential Input Peak-to-peak Voltage	Operational			1.2	V <sub>ppd</sub>	
	Input source common-mode noise	DC – 200MHz			150	mV <sub>ppd</sub>	
T <sub>TX-IDLE-SET-TO-</sub> IDLE	Max time to electrical idle after sending an EIOS			4	8	ns	
T <sub>TX-IDLE-TO-DIFF-</sub> DATA	Max time to valid diff signal after leaving electrical idle			4	8	ns	
V <sub>th+</sub>	On threshold of signal detector	Signal swing @ 4GHz	130		210	mV <sub>ppd</sub>	
V <sub>th</sub> -	Off threshold of signal detector	Signal swing @ 100MHz	30		110	mV <sub>ppd</sub>	
$V_{ m DD}$	Power supply voltage		3	3.3	3.6	V	
P <sub>max</sub>	Max Supply power	PRSNT#=0			1	W	
I <sub>max</sub>	Max Supply current				330	mA	
P <sub>idle</sub>	Supply power	PRSNT#=1			36	mW	
G <sub>P5GHZ</sub>	Peaking gain (Compensation at 5GHz, relative to 100MHz, 100mVp-p sine wave input) SW<1:0>=01, FG<1:0>=10	EQ<3:0> = 1111 EQ<3:0> = 1000 EQ<3:0> = 0000		16.1 13.5 8.0		dB	
	01, 10, 110, 110, 110, 110, 110, 110, 1	Variation around typical	-3		+3		
G <sub>P6GHZ</sub>	Peaking gain (Compensation at 6GHz, relative to 100MHz, 100mVp-p sine wave input) SW<1:0>=01, FG<1:0>=10	EQ<3:0> = 1111 EQ<3:0> = 1000 EQ<3:0> = 0000		17.2 14.8 9.0		dB	
		Variation around typical	-3		+3		
$G_{ m F}$	Flat gain (100MHz, EQ<3:0> = 1000, SW<1:0> = 01)	FG<1:0> = 11 FG<1:0> = 10 FG<1:0> = 01 FG<1:0> = 00		+2.0 -0.5 -2.0 -4.0		dB	
		Variation around typical	-3		+3		
$V_{1dB\_100M}$	-1dB compression point of output swing (at 100MHz)	SW0=1 SW0=0		1000 900		mV <sub>ppd</sub>	
V <sub>1dB_6</sub> G	-1dB compression point of output swing (at 6GHz) FG= 0dB, EQ = 0000 or 0(h)	SW0=1 SW0=0		600 540		mV <sub>ppd</sub>	





# High Speed I/O AC/DC Specifications Cont.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
V <sub>Coup</sub>	Channel isolation	100MHz to 5GHz, Figure 1 (Note 1)		28		dB
V	Input referred noise	100MHz to 5GHz, FG<1:0> = 11, EQ<3:0> = 0000, Figure 2		0.5		
V <sub>noise_input</sub>	Input-referred noise	100MHz to 5GHz, FG<1:0> = 11, EQ<3:0> = 1010, Figure 2		0.4		mV <sub>RMS</sub>
V	Output informal maior (Nets 2)	100MHz to 5GHz, FG<1:0> = 11, EQ<3:0> = 0000, Figure 2		0.7		
V <sub>noise_output</sub>	Output-referred noise (Note 2)	100MHz to 5GHz, FG<1:0> = 11, EQ<3:0> = 1010, Figure 2		0.8	1.6	mV <sub>RMS</sub>
Latency	·					
t <sub>pd</sub>	Latency	From input to output		0.2		ns
Jitter						
		PRBS31@24hrs 36" 5mils FR4				
R <sub>j</sub>	Additive Random Jitter at 8Gb/s (worst case)	VID = 0.8mVp-p $DE = 0dB$		0.0258		UI
		EQ = 0100				

Note 1: Measured using a vector-network analyzer (VNA) with -15dBm power level applied to the adjacent input. The VNA detects the signal at the output of the victim channel. All other inputs and outputs are terminated with  $50\Omega$ .

Note 2: Guaranteed by design and characterization.





# Characteristics of the SDA and SCI bus lines for Standard Mode I<sup>2</sup>C-bus devices<sup>(1)</sup>

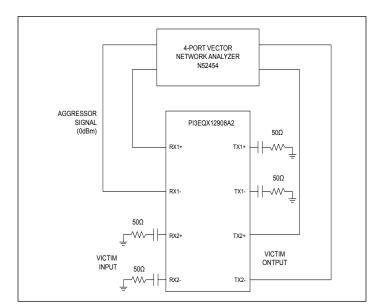
Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
$f_{SCL}$	SCL clock frequency			100	_	kHz
t <sub>HD;STA</sub>	Hold time (repeated) START condition. After this period, the first clock pulse is generated.		4.0		-	
t <sub>LOW</sub>	LOW period of the SCL clock		4.7		_	μs
t <sub>HIGH</sub>	HIGH period of the SCL clock		4.0		_	
t <sub>SU;STA</sub>	Set-up time for a repeated START condition		4.7		_	
t <sub>HD;DAT</sub>	Data hold time		10		_	ns
t <sub>SU;DAT</sub>	Data set-up time		250		_	
t <sub>r</sub>	Rise time of both SDA and SCL signals		-		1000	ns
t <sub>f</sub>	Fall time of both SDA and SCL signals				300	
t <sub>SU;STO</sub>	Set-up time for STOP condition		4.0		_	
t <sub>BUF</sub>	Bus free time between a STOP and STOP condition		4.7		_	μs
C <sub>b</sub>	Capacitive load for each bus line		-		400	pF

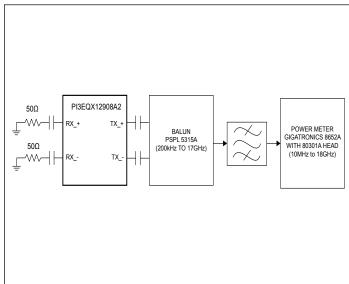
Notes:

<sup>1.</sup> All values referred to VIH min and VIL max levels



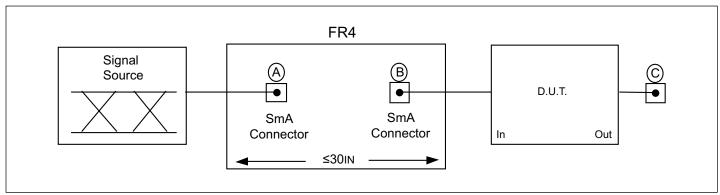






**Channel-Isolation Test Configuration** 

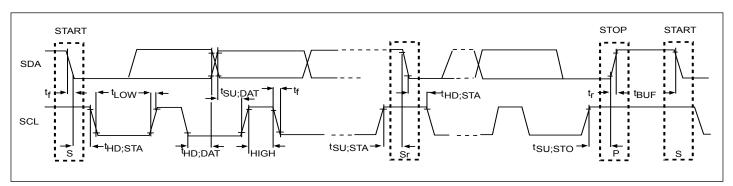
**Noise Test Configuration** 



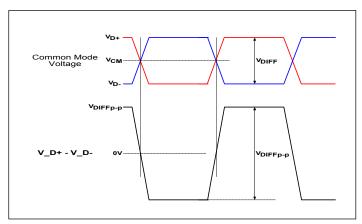
AC Test Circuit Referenced in the Electrical Characteristic Table







I<sup>2</sup>C Timing

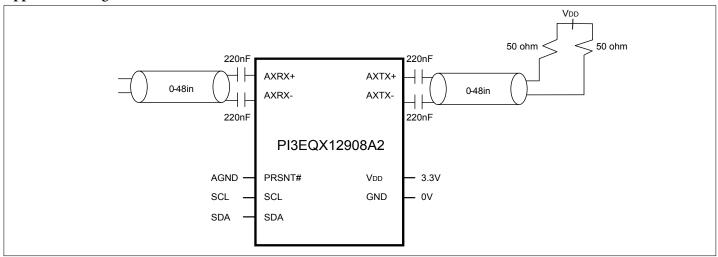


Definition of Differential Voltage and Differential Voltage Peak-to-Peak





#### **Application Diagrams**



# **Applications Information**

#### GENERAL RECOMMENDATIONS

The PI3EQX12908A2 is a high performance circuit capable of delivering excellent performance. Careful attention must be paid to the details associated with high-speed design as well as providing a clean power supply. Refer to the information below and Revision 4 of the LVDS Owner's Manual for more detailed information on high speed design tips to address signal integrity design issues.

#### PCB LAYOUT CONSIDERATIONS FOR DIFFERENTIAL PAIRS

The Differential inputs and LPDS outputs have been optimized to work with interconnects using a controlled differential impedance of 85 -  $100\Omega$ . It is preferable to route differential lines exclusively on one layer of the board, particularly for the input traces. The use of vias should be avoided if possible. If vias must be used, they should be used sparingly and must be placed symmetrically for each side of a given differential pair. Whenever differential vias are used the layout must also provide for a low inductance path for the return currents as well. Route the differential signals away from other signals and noise sources on the printed circuit board.

#### POWER SUPPLY BYPASSING

Two approaches are recommended to ensure that the PI3EQX12908A2 is provided with an adequate power supply. First, the supply (VDD) and ground (GND) pins should be connected to power planes routed on adjacent layers of the printed circuit board. The layer thickness of the dielectric should be minimized so that the VDD and GND planes create a low inductance supply with distributed capacitance. Second, careful attention to supply bypassing through the proper use of bypass capacitors is required. A 0.1 μF bypass capacitor should be connected to each VDD pin such that the capacitor is placed as close as possible to the PI3EQX12908A2. Smaller body size capacitors can help facilitate proper component placement. Additionally, capacitor with capacitance in the range of 1 µF to 10 μF should be incorporated in the power supply bypassing design as well. These capacitors can be either tantalum or an ultra-low ESR ceramic.

23

#### **Notes:**

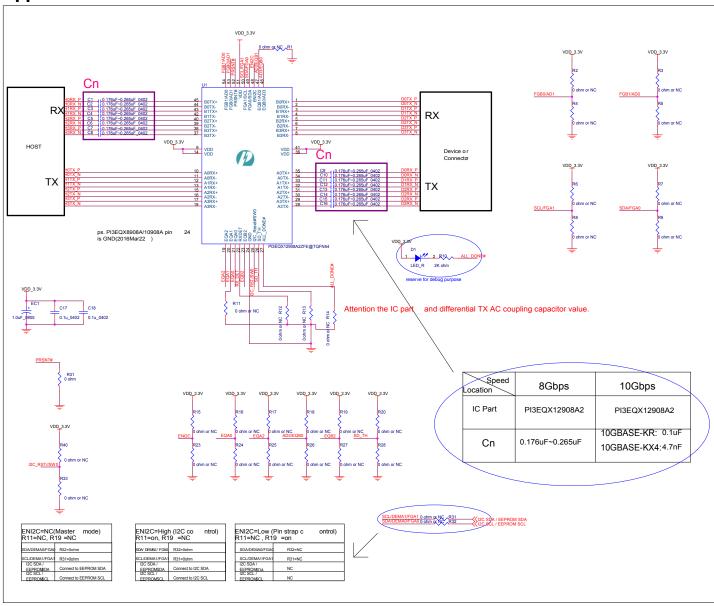
Hot Plug Detect feature operation is dependent on certain channel conditions, such as length.

For hot plug detect, reset will automatically go back to receiver detect (RXDET) cycle.





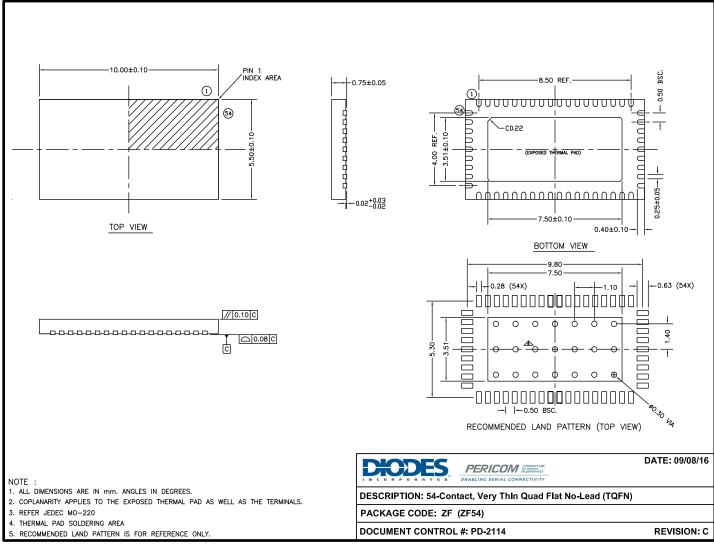
# **Application Schematics**

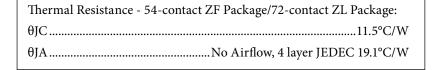






## Packaging Information 54-TQFN (ZF)





#### For latest package info.

 $please\ check: http://www.diodes.com/design/support/packaging/pericom-packaging/packaging-mechanicals-and-thermal-characteristics/pericom-packaging/packaging-mechanicals-and-thermal-characteristics/pericom-packaging-packaging-mechanicals-and-thermal-characteristics/pericom-packaging-$ 

## **Ordering Information**

Ordering Number	Package Code	Package Description
PI3EQX12908A2ZFEX	ZF	54-Contact, Very Thin Quad Flat No-Lead (TQFN)

#### Notes:

- · Thermal characteristics can be found on the company web site at www.diodes.com/design/support/packaging/
- E = Pb-free and Green
- X suffix = Tape/Reel





#### IMPORTANT NOTICE

DIODES INCORPORATED MAKES NO WARRANTY OF ANY KIND, EXPRESS OR IMPLIED, WITH REGARDS TO THIS DOCUMENT, INCLUDING, BUT NOT LIMITED TO, THE IMPLIED WARRANTIES OF MERCHANTABILITY AND FITNESS FOR A PARTICULAR PURPOSE (AND THEIR EQUIVALENTS UNDER THE LAWS OF ANY JURISDICTION).

Diodes Incorporated and its subsidiaries reserve the right to make modifications, enhancements, improvements, corrections or other changes without further notice to this document and any product described herein. Diodes Incorporated does not assume any liability arising out of the application or use of this document or any product described herein; neither does Diodes Incorporated convey any license under its patent or trademark rights, nor the rights of others. Any Customer or user of this document or products described herein in such applications shall assume all risks of such use and will agree to hold Diodes Incorporated and all the companies whose products are represented on Diodes Incorporated website, harmless against all damages.

Diodes Incorporated does not warrant or accept any liability whatsoever in respect of any products purchased through unauthorized sales channel.

Should Customers purchase or use Diodes Incorporated products for any unintended or unauthorized application, Customers shall indemnify and hold Diodes Incorporated and its representatives harmless against all claims, damages, expenses, and attorney fees arising out of, directly or indirectly, any claim of personal injury or death associated with such unintended or unauthorized application.

Products described herein may be covered by one or more United States, international or foreign patents pending. Product names and markings noted herein may also be covered by one or more United States, international or foreign trademarks.

This document is written in English but may be translated into multiple languages for reference. Only the English version of this document is the final and determinative format released by Diodes Incorporated.

#### LIFE SUPPORT

Diodes Incorporated products are specifically not authorized for use as critical components in life support devices or systems without the express written approval of the Chief Executive Officer of Diodes Incorporated. As used herein:

- A. Life support devices or systems are devices or systems which:
  - 1. are intended to implant into the body, or
- 2. support or sustain life and whose failure to perform when properly used in accordance with instructions for use provided in the labeling can be reasonably expected to result in significant injury to the user.
- B. A critical component is any component in a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or to affect its safety or effectiveness.

Customers represent that they have all necessary expertise in the safety and regulatory ramifications of their life support devices or systems, and acknowledge and agree that they are solely responsible for all legal, regulatory and safety-related requirements concerning their products and any use of Diodes Incorporated products in such safety-critical, life support devices or systems, notwithstanding any devices- or systems-related information or support that may be provided by Diodes Incorporated. Further, Customers must fully indemnify Diodes Incorporated and its representatives against any damages arising out of the use of Diodes Incorporated products in such safety-critical, life support devices or systems.

Copyright © 2016, Diodes Incorporated www.diodes.com